



## CAN TRANSCEIVER

### FEATURES

- Drop-In Improved Replacement for the PCA82C250 and PCA82C251
- Bus-Fault Protection of  $\pm 36$  V
- Meets or Exceeds ISO 11898
- Signaling Rates<sup>(1)</sup> up to 1 Mbps
- High Input Impedance Allows up to 120 SN65HVD251 Nodes on a Bus
- Bus Pin ESD Protection Exceeds 14 kV HBM
- Unpowered Node Does Not Disturb the Bus
- Low Current Standby Mode . . . 200  $\mu$ A Typical
- Thermal Shutdown Protection

- Glitch-Free Power-Up and Power-Down Bus Protection For Hot-Plugging
- DeviceNet Vendor ID # 806

### APPLICATIONS

- CAN Data Buses
- Industrial Automation
  - DeviceNet™ Data Buses
  - Smart Distributed Systems (SDS™)
- SAE J1939 Standard Data Bus Interface
- NMEA 2000 Standard Data Bus Interface
- ISO 11783 Standard Data Bus Interface

### DESCRIPTION

The SN65HVD251 is intended for use in applications employing the Controller Area Network (CAN) serial communication physical layer in accordance with the ISO 11898 Standard. The SN65HVD251 provides differential transmit capability to the bus and differential receive capability to a CAN controller at speeds up to 1 megabits per second (Mbps).

Designed for operation in harsh environments, the device features cross-wire, over-voltage and loss of ground protection to  $\pm 36$  V. Also featured are over-temperature protection as well as -7 V to 12 V common-mode range, and tolerance to transients of  $\pm 200$  V. The transceiver interfaces the single-ended CAN controller with the differential CAN bus found in industrial, building automation, and automotive applications.

Rs, pin 8, provides for three different modes of operation: high-speed, slope control, or low-power mode. The high-speed mode of operation is selected by connecting pin 8 to ground, allowing the transmitter output transistors to switch as fast as possible with no limitation on the rise and fall slope. The rise and fall slope can be adjusted by connecting a resistor to ground at pin 8; the slope is proportional to the pin's output current. Slope control with an external resistor value of 10 k $\Omega$  gives  $\sim 15$  V/us slew rate; 100 k $\Omega$  gives  $\sim 2$  V/us slew rate.

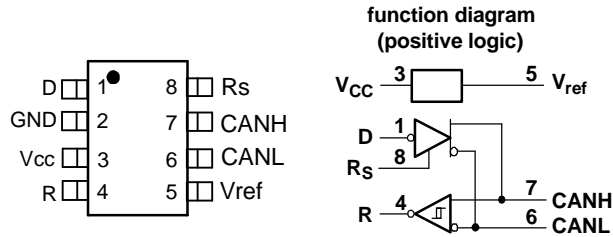
If a high logic level is applied to the Rs pin 8, the device enters a low-current standby mode during which the driver is switched off and the receiver remains active. The local protocol controller reverses this low-current standby mode when it needs to transmit to the bus.

The SN65HVD251 may be used in CAN, DeviceNet™ or SDS™ applications with the Texas Instruments' TMS320F241 and TMS320F243 DSPs with CAN 2.0B controllers.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

DeviceNet is a trademark of Allen-Bradley.  
SDS is a trademark of Honeywell.



(1) ORDERING INFORMATION

PART NUMBER	PACKAGE	MARKED AS
SN65HVD251D	8-pin SOIC (Tube)	VP251
SN65HVD251DR	8-pin SOIC (Tape & Reel)	VP251
SN65HVD251P	8-pin DIP	65HVD251

(1) (1)The signaling rate of a line is the number of voltage transitions that are made per second expressed in the units bps (bits per second).

ABSOLUTE MAXIMUM RATINGS (1), (2)

			SN65HVD251
Supply voltage range, $V_{CC}$			-0.3 V to 7V
Voltage range at any bus terminal (CANH or CANL)			-36 V to 36 V
Transient voltage per ISO 7637, pulse 1, 2, 3a, 3b			CANH, CANL $\pm 200$ V
Input voltage range, $V_I$ (D, Rs, or R)			-0.3 V to $V_{CC} + 0.5$
Electrostatic discharge	Human Body Model (3)	CANH, CANL and GND	14 kV
		All pins	6 kV
	Charged-Device Model (4)	All pins	1 kV
Continuous total power dissipation			(see Dissipation Rating Table)
Storage temperature range, $T_{stg}$			-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds			260°C

- (1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values, except differential I/O bus voltages, are with respect to network ground terminal.
- (3) Tested in accordance with JEDEC Standard 22, Test Method A114-A.
- (4) Tested in accordance with JEDEC Standard 22, Test Method C101.

ABSOLUTE MAXIMUM POWER DISSIPATION RATINGS

PACKAGE	CIRCUIT BOARD MODEL	$T_A = 25^\circ\text{C}$ POWER RATING	DERATING FACTOR (1) ABOVE $T_A = 25^\circ\text{C}$	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
SOIC (D)	Low-K(2)	600 mW	4.4 mW/°C	312 mW	120 mW
	High-K(3)	963 mW	7.7 mW/°C	501 mW	193 mW
PDIP (P)	Low-K(2)	984 mW	7.8 mW/°C	512 mW	197 mW
	High-K(3)	1344 mW	10.8 mW/°C	699 mW	269 mW

- (1) This is the inverse of the junction-to-ambient thermal resistance when board-mounted and with no air flow.
- (2) In accordance with the Low-K thermal metric definitions of EIA/JESD51-3.
- (3) In accordance with the High-K thermal metric definitions of EIA/JESD51-7.

THERMAL CHARACTERISTICS

PARAMETER	TEST CONDITIONS	VALUE			UNITS
		MIN	TYP	MAX	
$\Theta_{JB}$ Junction-to-board thermal resistance	D		78.7		°C/W
	P		56.5		

**THERMAL CHARACTERISTICS (continued)**

PARAMETER	TEST CONDITIONS	VALUE			UNITS
		MIN	TYP	MAX	
$\Theta_{JC}$ Junction-to-board thermal resistance	D		44.6		°C/W
	P		54.5		
$P_D$ Device power dissipation	$V_{CC} = 5\text{ V}$ , $T_j = 27\text{ °C}$ , $R_L = 60\Omega$ , $R_S$ at 0 V, Input to D a 500-kHz 50% duty cycle square wave			97.7	mW
	$V_{CC} = 5.5\text{ V}$ , $T_j = 130\text{ °C}$ , $R_L = 60\Omega$ , $R_S$ at 0 V, Input to D a 500-kHz 50% duty cycle square wave			142	mW
$T_{SD}$ Thermal shutdown junction temperature			165		°C

**RECOMMENDED OPERATING CONDITIONS**

over recommended operating conditions (unless otherwise noted).

PARAMETER		MIN	NOM	MAX	UNIT
Supply voltage, $V_{CC}$		4.5		5.5	V
Voltage at any bus terminal (separately or common mode) $V_I$ or $V_{IC}$		-7 <sup>(1)</sup>		12	V
High-level input voltage, $V_{IH}$	D input	0.7 $V_{CC}$			V
Low-level input voltage, $V_{IL}$	D input			0.3 $V_{CC}$	V
Differential input voltage, $V_{ID}$		-6		6	V
Input voltage to $R_S$ , $V_{I(R_S)}$		0		$V_{CC}$	V
Input voltage at $R_S$ for standby, $V_{I(R_S)}$		0.75 $V_{CC}$		$V_{CC}$	V
$R_S$ wave-shaping resistance		0		100	k $\Omega$
High-level output current, $I_{OH}$	Driver	-50			mA
	Receiver	-4			
Low-level output current, $I_{OL}$	Driver			50	mA
	Receiver			4	
Operating free-air temperature, $T_A$		-40		125	°C
Junction temperature, $T_j$	PDIP Package			145	°C
	SOIC Package			150	

(1) The algebraic convention, in which the least positive (most negative) limit is designated as minimum is used in this data sheet.

**DRIVER ELECTRICAL CHARACTERISTICS**

over recommended operating conditions (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP <sup>(1)</sup>	MAX	UNIT
$V_{O(D)}$ Bus output voltage (Dominant)	CANH	Figure 1 & Figure 2 , D at 0 V $R_S$ at 0 V	2.75	3.5	4.5	V
	CANL		0.5		2	
$V_{O(R)}$ Bus output voltage (Recessive)	CANH	Figure 1 & Figure 2 , D at 0.7 $V_{CC}$ , $R_S$ at 0 V	2	2.5	3	V
	CANL		2	2.5	3	
$V_{OD(D)}$ Differential output voltage (Dominant)		Figure 1 , D at 0 V, $R_S$ at 0 V	1.5	2	3	V
$V_{OD(D)}$ Differential output voltage (Dominant)		Figure 2 & Figure 3 , D at 0 V, $R_S$ at 0 V	1.2	2	3.1	V
$V_{OD(R)}$ Differential output voltage (Recessive)		Figure 1 & Figure 2 , D at 0.7 $V_{CC}$	-120		12	mV
$V_{OD(R)}$ Differential output voltage (Recessive)		D at 0.7 $V_{CC}$ , no load	-0.5		0.05	V
$V_{OC(pp)}$ Peak-to-peak common-mode output voltage		Figure 9, $R_S$ at 0 V		600		mV
$I_{IH}$ High-level input current, D Input		D at 0.7 $V_{CC}$	-40		0	$\mu$ A
$I_{IL}$ Low-level input current, D Input		D at 0.3 $V_{CC}$	-60		0	$\mu$ A

(1) All typical values are at 25°C and with a 5-V supply.

**DRIVER ELECTRICAL CHARACTERISTICS (continued)**

over recommended operating conditions (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP <sup>(1)</sup>	MAX	UNIT
$I_{OS(SS)}$	Short-circuit steady-state output current	Figure 11, $V_{CANH}$ at -7 V, CANL Open	-200			mA
		Figure 11, $V_{CANH}$ at 12 V, CANL Open			2.5	
		Figure 11, $V_{CANL}$ at -7 V, CANH Open	-2			
		Figure 11, $V_{CANL}$ at 12 V, CANH Open			200	
$C_O$	Output capacitance	See receiver input capacitance				
$I_{OZ}$	High-impedance output current	See receiver input current				
$I_{IRs(s)}$	$R_s$ input current for standby	$R_s$ at $0.75 V_{CC}$	-10			$\mu A$
$I_{IRs(f)}$	$R_s$ input current for full speed operation	$R_s$ at 0 V	-550		0	$\mu A$
$I_{CC}$	Supply current	Standby	$R_s$ at $V_{CC}$ , D at $V_{CC}$		275	$\mu A$
		Dominant	D at 0 V, $60\Omega$ load, $R_s$ at 0 V		65	mA
		Recessive	D at $V_{CC}$ , no load, $R_s$ at 0 V		14	

**DRIVER SWITCHING CHARACTERISTICS**

over recommended operating conditions (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
$t_{pLH}$	Propagation delay time, low-to-high-level output	Figure 4, $R_s$ at 0 V		40	70	ns	
		Figure 4, $R_s$ with 10 k $\Omega$ to ground		90	125		
		Figure 4, $R_s$ with 100 k $\Omega$ to ground		500	800		
$t_{pHL}$	Propagation delay time, high-to-low-level output	Figure 4, $R_s$ at 0 V		85	125		
		Figure 4, $R_s$ with 10 k $\Omega$ to ground		200	260		
		Figure 4, $R_s$ with 100 k $\Omega$ to ground		1150	1450		
$t_{sk(p)}$	Pulse skew ( $ t_{pHL} - t_{pLH} $ )	Figure 4, $R_s$ at 0 V		45	85		
		Figure 4, $R_s$ with 10 k $\Omega$ to ground		110	180		
		Figure 4, $R_s$ with 100 k $\Omega$ to ground		650	900		
$t_r$	Differential output signal rise time	Figure 4, $R_s$ at 0 V	35		100	ns	
$t_f$	Differential output signal fall time		35		100		
$t_r$	Differential output signal rise time	Figure 4, $R_s$ with 10 k $\Omega$ to ground	100		250		
$t_f$	Differential output signal fall time		100		250		
$t_r$	Differential output signal rise time	Figure 4, $R_s$ with 100 k $\Omega$ to ground	600		1550		
$t_f$	Differential output signal fall time		600		1550		
$t_{en}$	Enable time from standby to dominant	Figure 8			0.5		$\mu s$

**RECEIVER ELECTRICAL CHARACTERISTICS**

over recommended operating conditions (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{IT+}$	Positive-going input threshold voltage	$R_s$ at 0 V, (See Table 1)		750	900	mV
$V_{IT-}$	Negative-going input threshold voltage		500	650		
$V_{hys}$	Hysteresis voltage ( $V_{IT+} - V_{IT-}$ )		100			
$V_{OH}$	High-level output voltage	Figure 6, $I_O = -4mA$	0.8 $V_{CC}$			V
$V_{OL}$	Low-level output voltage	Figure 6, $I_O = 4mA$		0.2 $V_{CC}$		V

## RECEIVER ELECTRICAL CHARACTERISTICS (continued)

over recommended operating conditions (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$I_I$	Bus input current	CANH or CANL at 12 V			600	$\mu\text{A}$
		CANH or CANL at 12 V, $V_{CC}$ at 0 V	Other bus pin at 0 V, $R_s$ at 0 V, D at $0.7 V_{CC}$		715	
		CANH or CANL at -7 V		-460		
		CANH or CANL at -7 V, $V_{CC}$ at 0 V		-340		
$C_I$	Input capacitance, (CANH or CANL)	Pin-to-ground, $V_I = 0.4 \sin(4E6\pi t) + 0.5 \text{ V}$ , D at $0.7 V_{CC}$		20		pF
$C_{ID}$	Differential input capacitance	Pin-to-pin, $V_I = 0.4 \sin(4E6\pi t) + 0.5 \text{ V}$ , D at $0.7 V_{CC}$		10		pF
$R_{ID}$	Differential input resistance	D at $0.7 V_{CC}$ , $R_s$ at 0 V	40		100	k $\Omega$
$R_{IN}$	Input resistance, (CANH or CANL)	D at $0.7 V_{CC}$ , $R_s$ at 0 V	20		50	k $\Omega$
$I_{CC}$	Supply current	Standby	$R_s$ at $V_{CC}$ , D at $V_{CC}$		275	$\mu\text{A}$
		Dominant	D at 0 V, 60 $\Omega$ Load, $R_s$ at 0 V		65	mA
		Recessive	D at $V_{CC}$ , No Load, $R_s$ at 0 V		14	

## RECEIVER SWITCHING CHARACTERISTICS

over recommended operating conditions (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{pLH}$	Propagation delay time, low-to-high-level output	Figure 6		35	50	ns
$t_{pHL}$	Propagation delay time, high-to-low-level output			35	50	
$t_{sk(p)}$	Pulse skew ( $t_{pHL} - t_{pLH}$ )				20	
$t_r$	Output signal rise time			2	4	
$t_f$	Output signal fall time			2	4	
$t_{p(sb)}$	Propagation delay time in standby	Figure 12, $R_s$ at $V_{CC}$			500	

## VREF-PIN CHARACTERISTICS

over recommended operating conditions (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_O$	Reference output voltage	$-5 \mu\text{A} < I_O < 5 \mu\text{A}$	$0.45 V_{CC}$		$0.55 V_{CC}$	V
		$-50 \mu\text{A} < I_O < 50 \mu\text{A}$	$0.4 V_{CC}$		$0.6 V_{CC}$	

**DEVICE SWITCHING CHARACTERISTICS**

over recommended operating conditions (unless otherwise noted).

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{loop1}$	Figure 10, $R_s$ at 0 V		60	100	ns
	Figure 10, $R_s$ with 10 k $\Omega$ to ground		100	150	
	Figure 10, $R_s$ with 100 k $\Omega$ to ground		440	800	
$t_{loop2}$	Figure 10, $R_s$ at 0 V		115	150	ns
	Figure 10, $R_s$ with 10 k $\Omega$ to ground		235	290	
	Figure 10, $R_s$ with 100 k $\Omega$ to ground		1070	1450	
$t_{loop2}$	Figure 10, $R_s$ at 0 V, $V_{CC}$ from 4.5 V to 5.1 V,		105	145	ns

**PARAMETER MEASUREMENT INFORMATION**

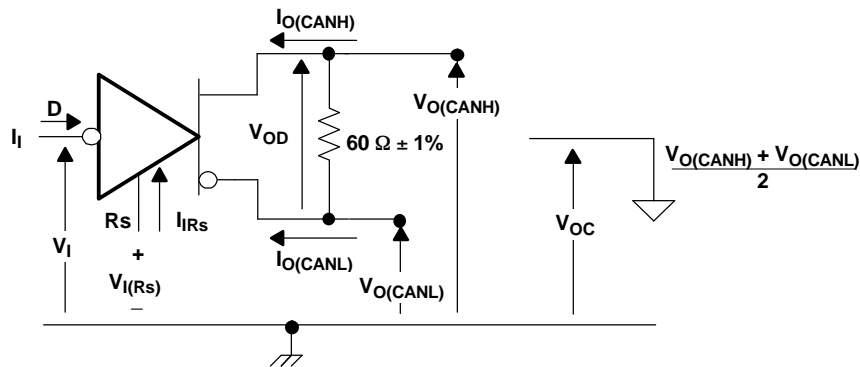


Figure 1. Driver Voltage, Current, and Test Definition

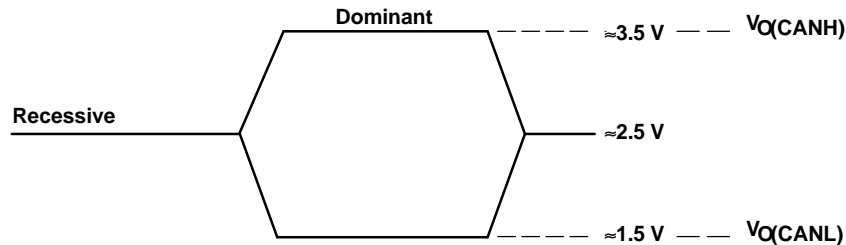


Figure 2. Bus Logic State Voltage Definitions

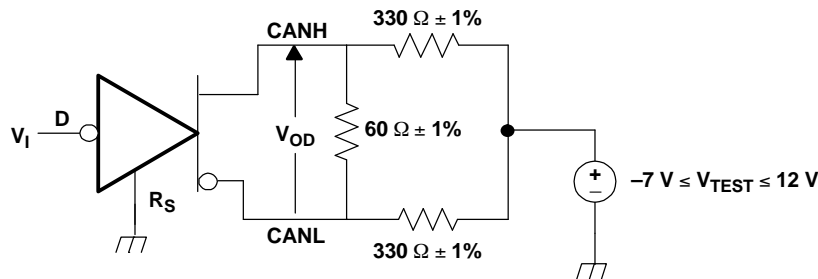


Figure 3. Driver  $V_{OD}$

PARAMETER MEASUREMENT INFORMATION (continued)

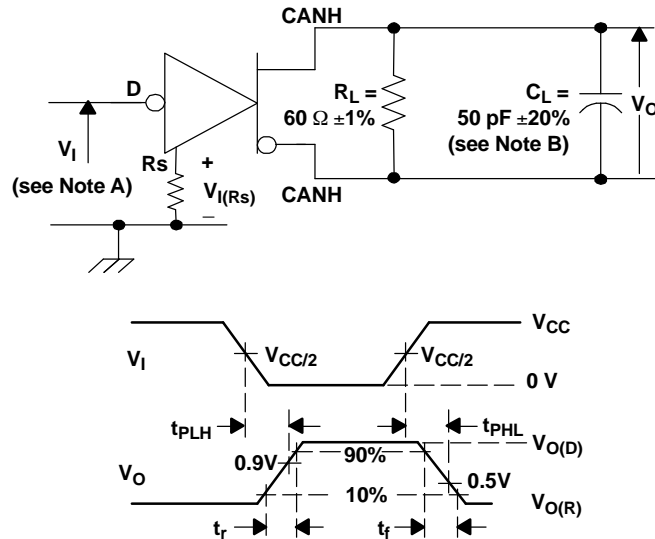


Figure 4. Driver Test Circuit and Voltage Waveforms

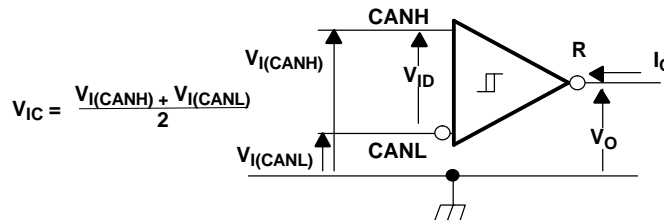
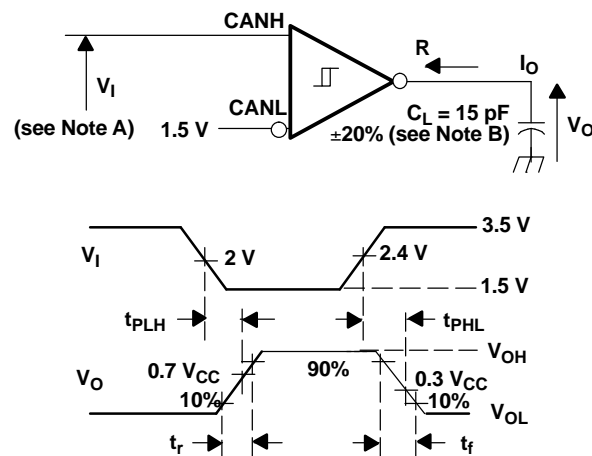


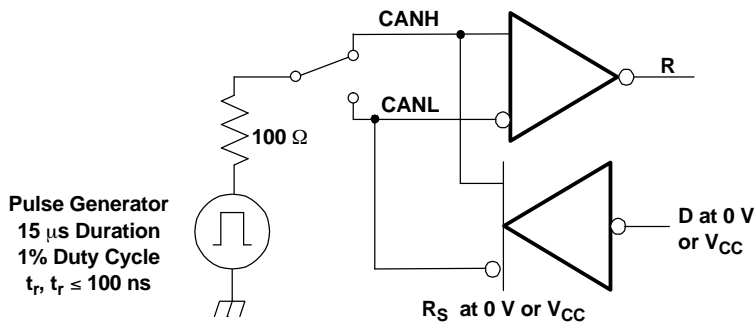
Figure 5. Receiver Voltage and Current Definitions



- A. The input pulse is supplied by a generator having the following characteristics: PRR ≤ 125 kHz, 50% duty cycle,  $t_r \leq 6\text{ns}$ ,  $t_f \leq 6\text{ns}$ ,  $Z_0 = 50\Omega$ .
- B.  $C_L$  includes instrumentation and fixture capacitance within  $\pm 20\%$ .

Figure 6. Receiver Test Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION (continued)



A. This test is conducted to test survivability only. Data stability at the R output is not specified.

Figure 7. Test Circuit, Transient Over-Voltage Test

Table 1. Receiver Characteristics Over Common Mode Voltage

INPUT		MEASURED	OUTPUT	
$V_{CANH}$	$V_{CANL}$	$ V_{ID} $	R	
12 V	11.1 V	900 mV	L	$V_{OL}$
-6.1 V	-7 V	900 mV	L	
-1 V	-7 V	6 V	L	
12 V	6 V	6 V	L	
-6.5 V	-7 V	500 mV	H	$V_{OH}$
12 V	11.5 V	500 mV	H	
-7 V	-1 V	6 V	H	
6 V	12 V	6 V	H	
open	open	X	H	

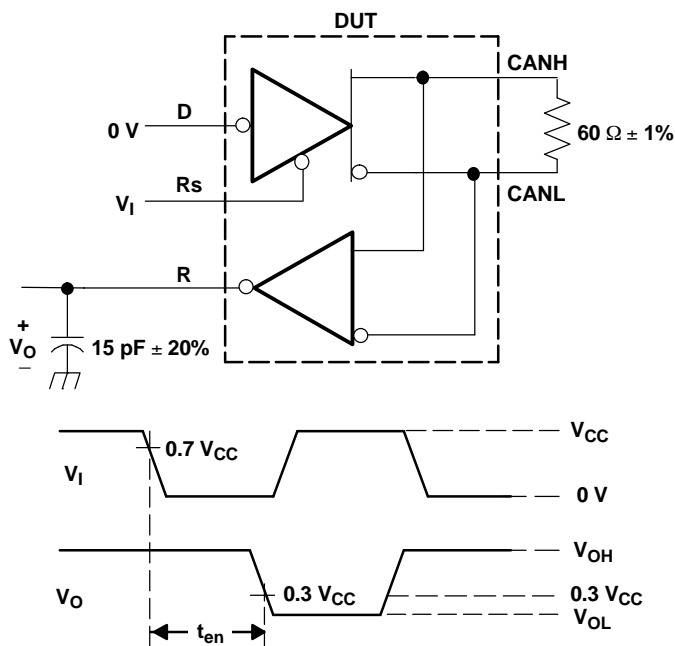
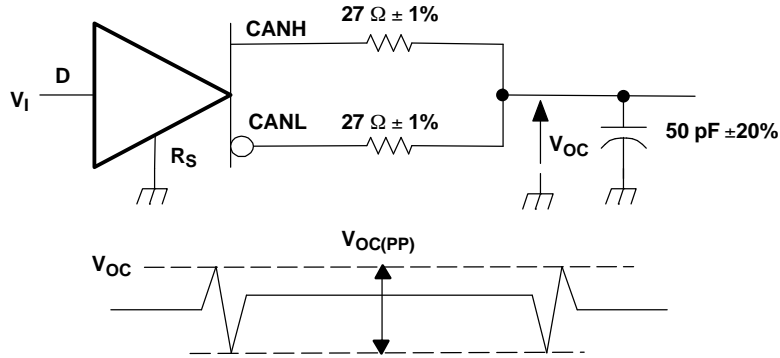


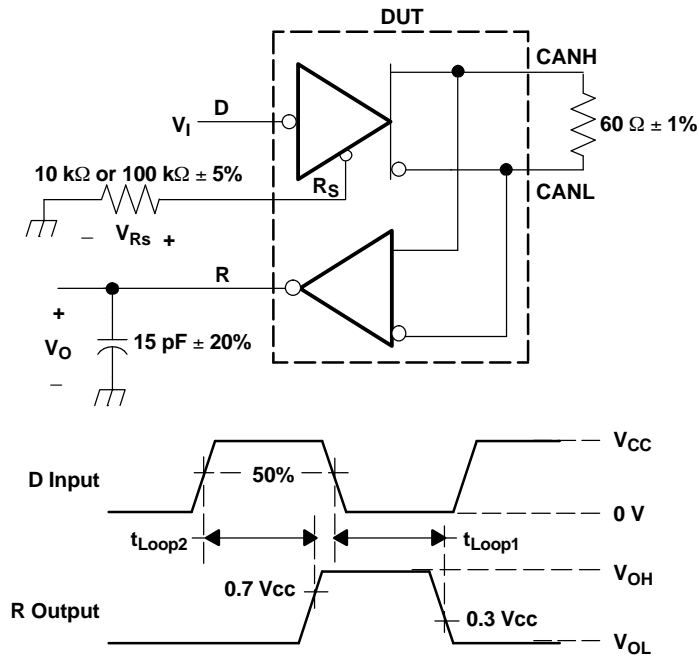
Figure 8.  $t_{en}$  Test Circuit and Voltage Waveforms





- A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  125 kHz, 50% duty cycle,  $t_r \leq$  6ns,  $t_f \leq$  6ns,  $Z_O = 50\Omega$ .

**Figure 9. Peak-to-Peak Common Mode Output Voltage**



**Figure 10.  $t_{LOOP}$  Test Circuit and Voltage Waveforms**

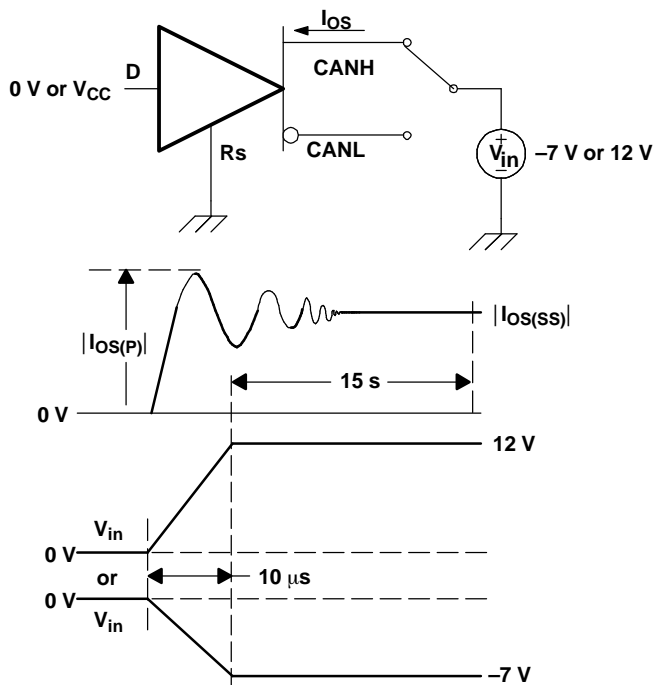
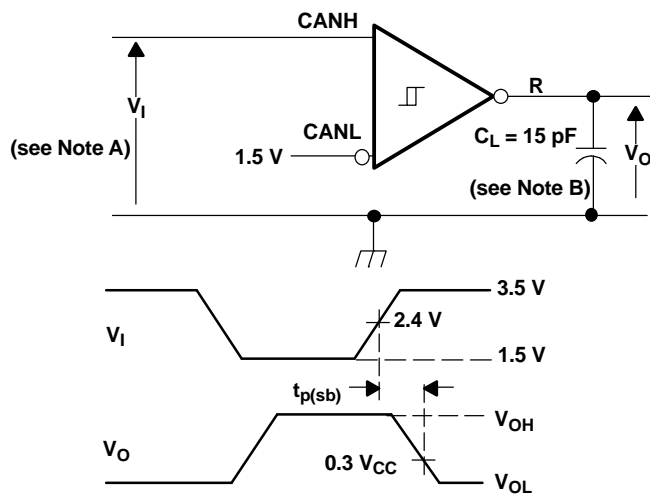


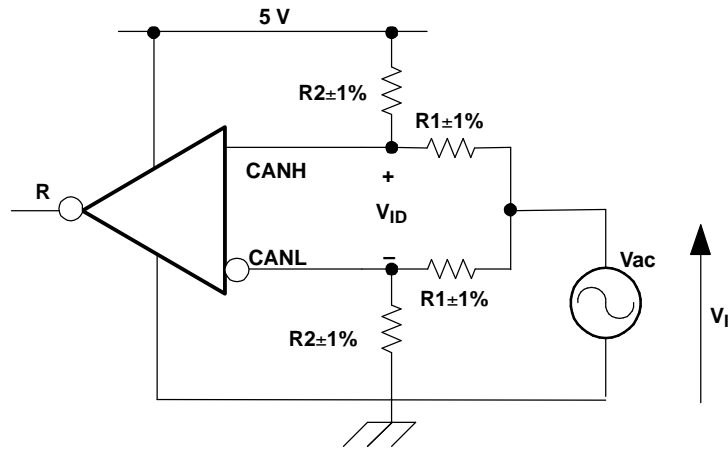
Figure 11. Driver Short-Circuit Test



- A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  125 kHz, 50% duty cycle,  $t_r \leq$  6ns,  $t_f \leq$  6ns,  $Z_0 = 50\Omega$ .
- B.  $C_L$  includes instrumentation and fixture capacitance within  $\pm 20\%$ .

Figure 12. Receiver Propagation Delay in Standby Test Circuit and Waveform

DEVICE INFORMATION



$V_{ID}$	R1	R2
500 mV	50 $\Omega$	450 $\Omega$
900 mV	50 $\Omega$	227 $\Omega$



A. All input pulses are supplied by a generator having the following characteristics:  $f < 1.5$  MHz,  $T_A = 25^\circ\text{C}$ ,  $V_{CC} = 5.0$  V.

Figure 13. Common-Mode Input Voltage Rejection Test

FUNCTION TABLES

Table 2. DRIVER

INPUTS	Voltage at $R_S$ , $V_{RS}$	OUTPUTS		BUS STATE
		CANH	CANL	
L	$V_{RS} < 1.2$ V	H	L	Dominant
H	$V_{RS} < 1.2$ V	Z	Z	Recessive
Open	X	Z	Z	Recessive
X	$V_{RS} > 0.75 V_{CC}$	Z	Z	Recessive

Table 3. RECEIVER

DIFFERENTIAL INPUTS [ $V_{ID} = V(\text{CANH}) - V(\text{CANL})$ ]	OUTPUT R <sup>(1)</sup>
$V_{ID} \geq 0.9$ V	L
$0.5\text{V} < V_{ID} < 0.9$ V	?
$V_{ID} \leq 0.5$ V	H
Open	H

(1) H = high level; L = low level; X = irrelevant; ? = indeterminate; Z = high impedance

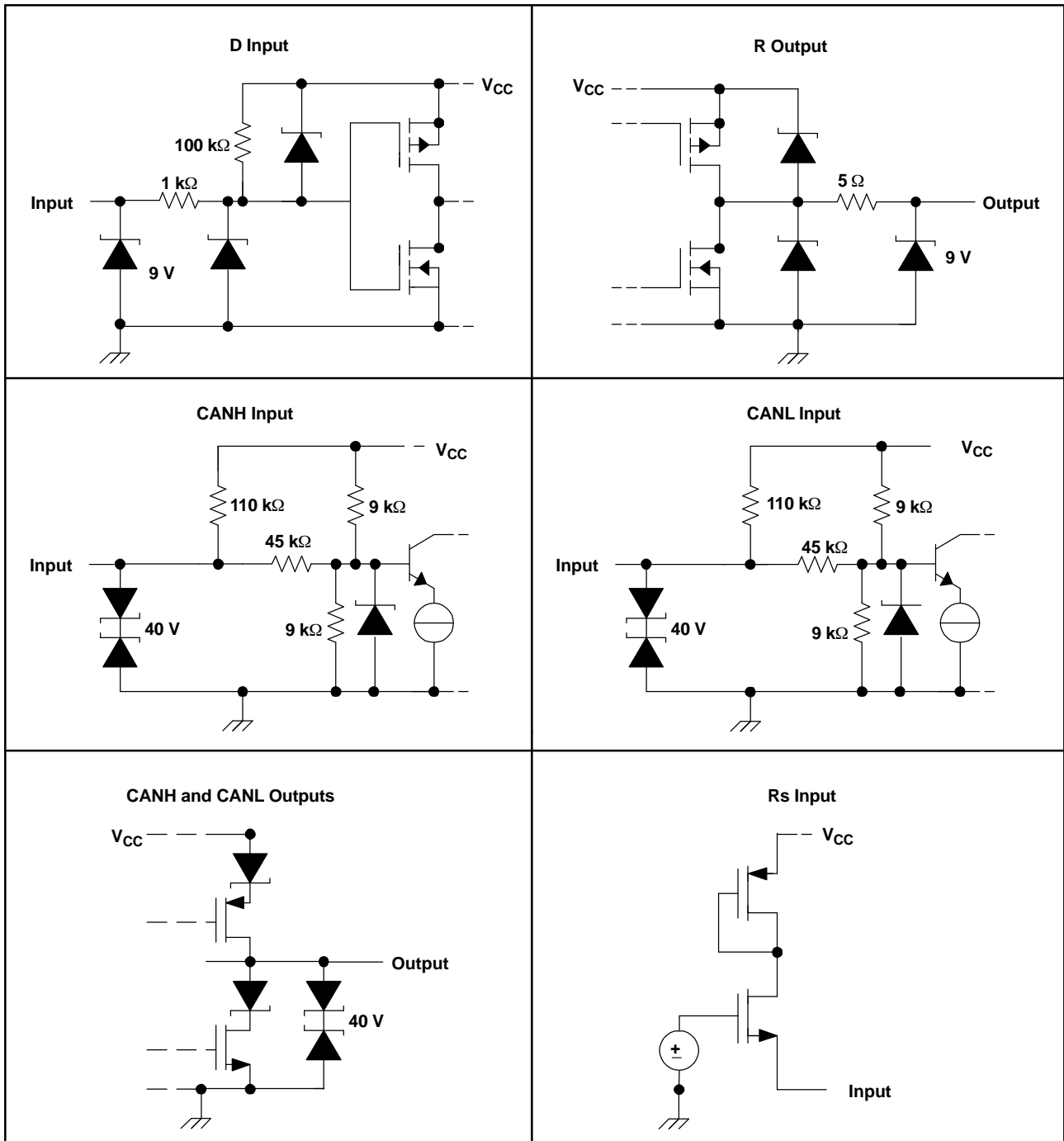
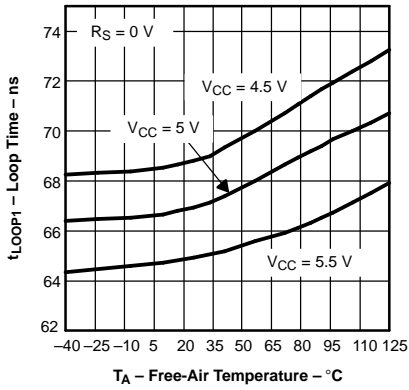


Figure 14. Equivalent Input and Output Schematic Diagrams

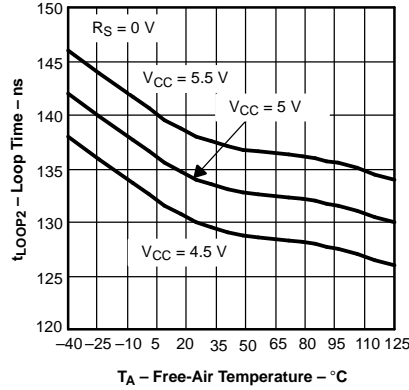
**TYPICAL CHARACTERISTICS**

**t<sub>LOOP1</sub>-LOOP TIME  
VS  
FREE-AIR TEMPERATURE**



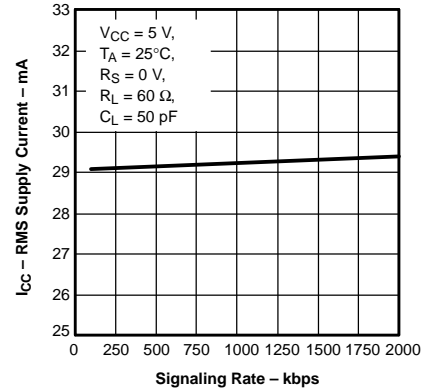
**Figure 15.**

**t<sub>LOOP2</sub>-LOOP TIME  
VS  
FREE-AIR TEMPERATURE**



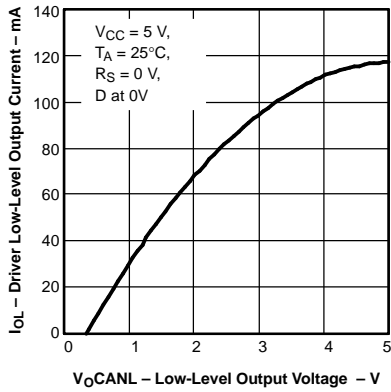
**Figure 16.**

**SUPPLY CURRENT (RMS)  
VS  
SIGNALING RATE**



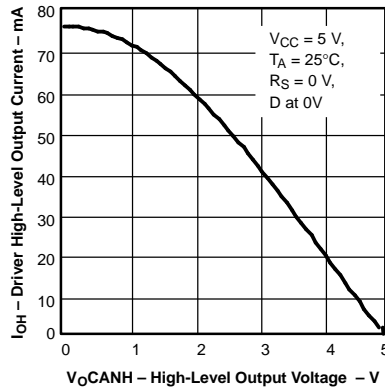
**Figure 17.**

**DRIVER LOW-LEVEL OUTPUT CURRENT  
VS  
LOW-LEVEL OUTPUT VOLTAGE**



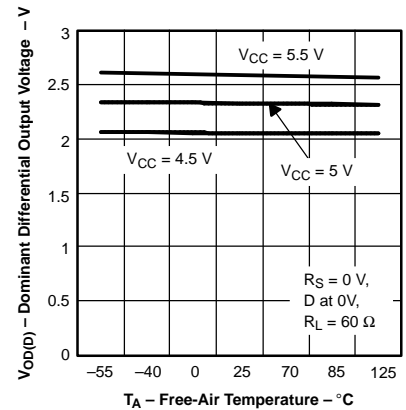
**Figure 18.**

**DRIVER HIGH-LEVEL OUTPUT CURRENT  
VS  
HIGH-LEVEL OUTPUT VOLTAGE**

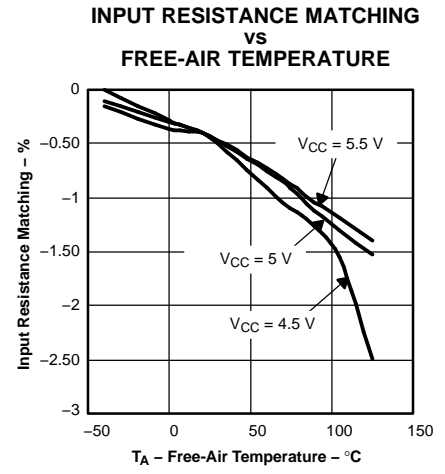
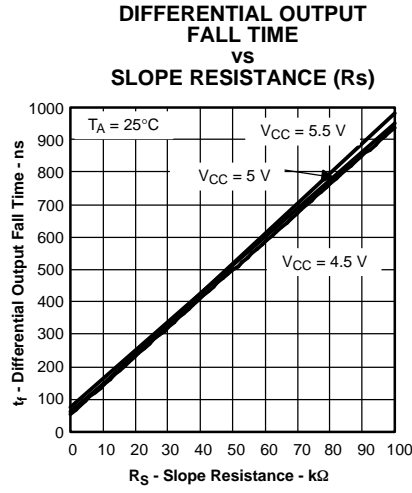
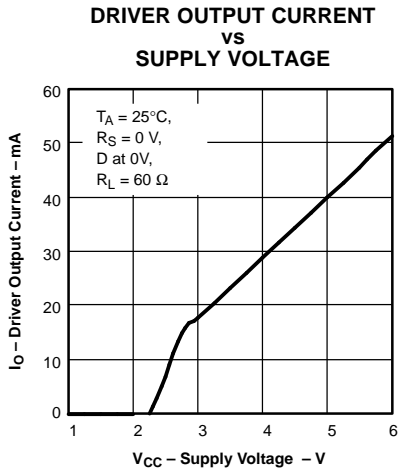


**Figure 19.**

**DOMINANT DIFFERENTIAL OUTPUT VOLTAGE  
VS  
FREE-AIR TEMPERATURE**



**Figure 20.**



## APPLICATION INFORMATION

The basics of arbitration require that the receiver at the sending node designate the first bit as dominant or recessive after the initial wave of the first bit of a message travels to the most remote node on a network and back again. Typically, this *sample* is made at 75% of the bit width, and within this limitation, the maximum allowable signal distortion in a CAN network is determined by network electrical parameters.

Factors to be considered in network design include the  $\approx 5$  ns/m propagation delay of typical twisted-pair bus cable; signal amplitude loss due to the loss mechanisms of the cable; and the number, length, and spacing of drop-lines (stubs) on a network. Under strict analysis, variations among the different oscil-

lators in a system also need to be accounted for with adjustments in signaling rate and stub & bus length. Table 2 lists the maximum signaling rates achieved with the SN65HVD251 in high-speed mode with several bus lengths of category 5, shielded twisted pair (CAT 5 STP) cable.

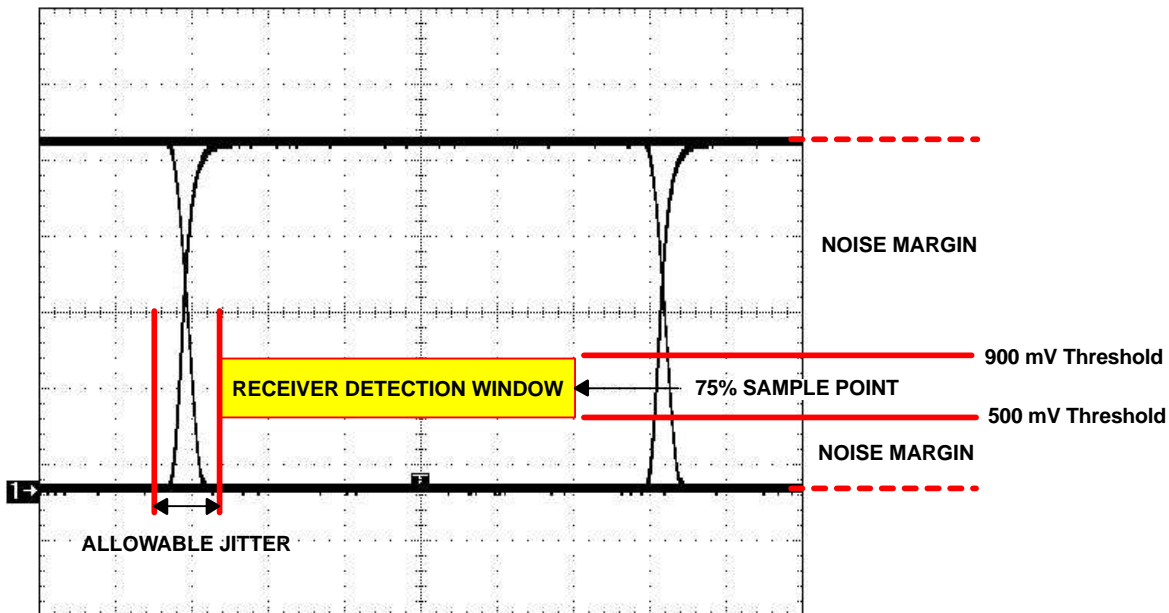
**Table 4. Maximum Signaling Rates for Various Cable Lengths**

BUS LENGTH (m)	SIGNALING RATE (kbps)
30	1000
100	500
250	250
500	125
1000	62.5

The ISO 11898 Standard specifies a maximum bus length of 40 m and maximum stub length of 0.3 m with a maximum of 30 nodes. However, with careful design, users can have longer cables, longer stub lengths, and many more nodes on a bus. (Note: Non-standard application may come with a trade-off in signaling rate.) A large number of nodes requires a transceiver with high input impedance such as the HVD251.

The Standard specifies the interconnect to be a single twisted-pair cable (shielded or unshielded) with  $120\Omega$  characteristic impedance ( $Z_0$ ). Resistors equal to the characteristic impedance of the line terminate both ends of the cable to prevent signal reflections. Unterminated drop-lines connect nodes to the bus and should be kept as short as possible to minimize signal reflections.

Connectors, while not specified by the Standard should have as little effect as possible on standard operating parameters such as capacitive loading. Although unshielded cable is used in many applications, data transmission circuits employing CAN transceivers are usually used in applications requiring a rugged interconnection with a wide common-mode voltage range. Therefore, shielded cable is recommended in these electronically harsh environments, and when coupled with the Standard's  $-2$ -V to  $7$ -V common-mode range of tolerable ground noise, helps to ensure data integrity. The HVD251 enhances the Standard's insurance of data integrity with an extended  $-7$ -V to  $12$ -V range of common-mode operation.



**Figure 24. Typical CAN Differential Signal Eye-Pattern**

An eye pattern is a useful tool for measuring overall signal quality. As displayed in Figure 24, the differential signal changes logic states in two places on the display, producing an eye. Instead of viewing only one logic crossing on the scope, an entire *bit* of data is brought into view. The resulting eye pattern includes all of the effects of systemic and random distortion, and displays the time during which a signal may be considered valid.

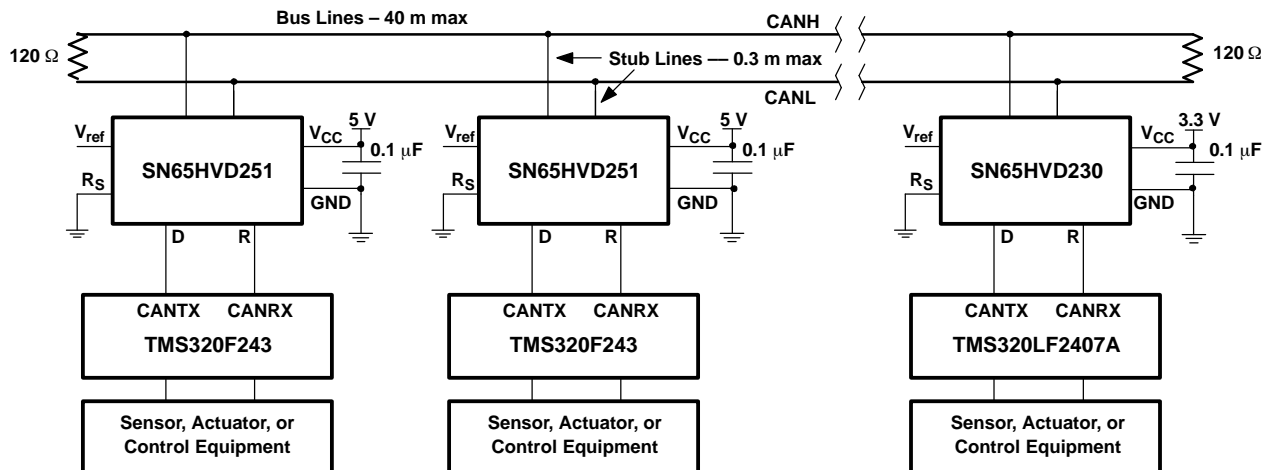
The height of the eye above or below the receiver threshold voltage level at the sampling point is the noise margin of the system. Jitter is typically measured at the differential voltage zero-crossing during the logic state transition of a signal. Note that jitter present at the receiver threshold voltage level is considered by some to be a more effective representation of the jitter at the input of a receiver.

As the sum of skew and noise increases, the eye closes and data is corrupted. Closing the width decreases the time available for accurate sampling, and lowering the height enters the 900 mV or 500 mV threshold of a receiver.

Different sources induce noise onto a signal. The more obvious noise sources are the components of a transmission circuit themselves; the signal transmitter, traces & cables, connectors, and the receiver. Beyond that, there is a termination dependency, cross-talk from clock traces and other proximity effects, VCC & ground bounce, and electromagnetic interference from near-by electrical equipment.

The balanced receiver inputs of the HVD251 mitigate most all sources of signal corruption, and when used with a quality shielded twisted-pair cable, help insure data integrity.

**Typical Application**



**Figure 25. Typical HVD251 Application**



P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE

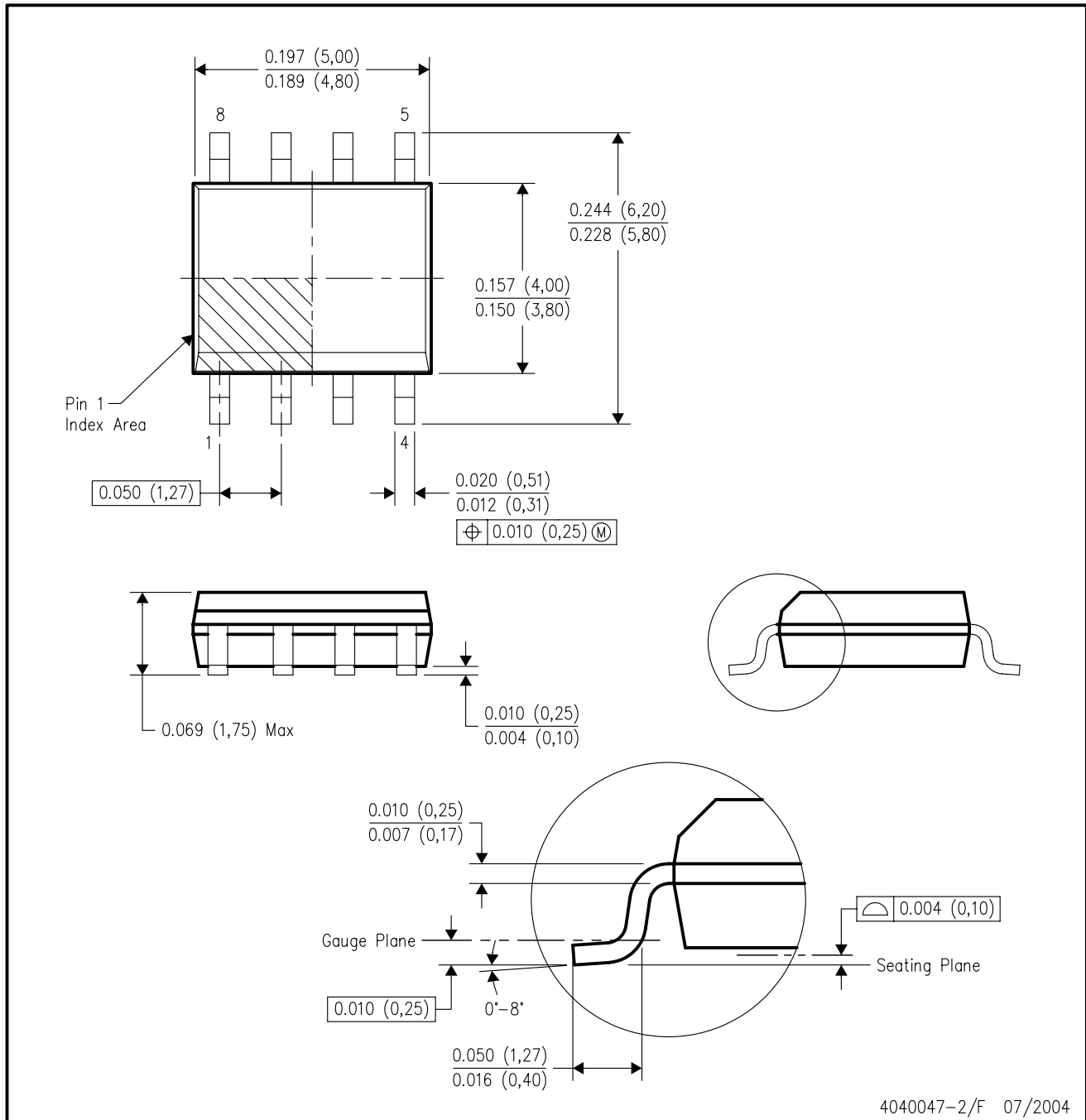


- NOTES: A. All linear dimensions are in inches (millimeters).  
 B. This drawing is subject to change without notice.  
 C. Falls within JEDEC MS-001

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D (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
  - B. This drawing is subject to change without notice.
  - C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
  - D. Falls within JEDEC MS-012 variation AA.

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